Three-dimensional imaging with 1.06 μm Geiger-mode LADAR camera

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ABSTRACT

Three-dimensional (3D) topographic imaging using Short wavelength infrared (SWIR) Laser Detection and Range (LADAR) systems have been successfully demonstrated on various platforms. LADAR imaging provides coverage down to inch-level fidelity and allows for effective wide-area terrain mapping. Recently Spectrolab has demonstrated a compact 32x32 LADAR camera with single photon-level sensitivity with small size, weight, and power (SWAP) budget. This camera has many special features such as non-uniform bias correction, variable range gate width from 2 microseconds to 6 microseconds, windowing for smaller arrays, and shorted pixel protection. Boeing integrated this camera with a 1.06 µm pulse laser on various platforms and had demonstrated 3D imaging. In this presentation, the operation details of this camera and 3D imaging demonstration using this camera on various platforms will be presented.

1. INTRODUCTION AND BACKGROUND

SWIR 3D imaging provides one more important dimension in foliage penetration, camouflage imaging, and aerial mapping in battlefield intelligence and earth survey. The potential market for its airborne applications is greatly influenced by the size, weight, and power (SWAP) of LADAR systems. Since the laser and its service system takes most of the system SWAP, it is essential to maximize the camera sensitivity to reduce the laser power in order to enhance the ranging distance and reduce the SWAP of the whole system. By reaching the optical detection quantum limit, InP-based single-photon counting Geiger-mode avalanche photodiodes (GM-APD) fit perfectly to this requirement. SWIR GM-APD focal plane arrays (GM-FPAs) have been reported by both MIT Lincoln Laboratory¹ and Boeing Spectrolab². Due to the availability of high power emitters, most of the effort to date has been focused on photodiodes operating at 1.06 µm.

Important figures of merits for GM-APDs include dark count rate (DCR) and photon detection efficiency (PDE) which together establish the upper limit of the signal-to-noise ratio for the camera. For active 3D imaging, the upper limit of range resolution is determined by FPA timing jitter, while the spatial resolution is largely determined by FPA crosstalk. Frame rate describes how often the camera can take a 3D shot. Because GM-APD cameras normally require multiple frames to get clear pictures, and this parameter will greatly influence the aircraft speed and survey efficiency. Since the frame rate is primarily determined by the data process and download rate, the GM-APD afterpulsing, or temporal crosstalk between range gates, is normally not a great concern in imaging applications as long as it is no less than 100 kHz. Other important parameters include weight, power, and volume. Because most of the power consumed in a

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LADAR camera is by the thermoelectric coolers (TEC), a higher APD operation temperature is highly desirable to airborne applications.

We have been improving the performance of both the avalanche photodiode detector and ROIC arrays that together comprise an FPA. The enhanced capabilities of the newly designed 32x32 ROIC include non-uniform bias (NUB) correction and short pixel protection. Instead of biasing the whole array with a single voltage in most focal plane arrays (FPA), NUB can tune the bias individually to pixels with 4-bit resolution and in a 2.5 V range. High voltage is applied to pixels to operate in Geiger mode. A single pixel shorting out of the thousands of pixels can reduce the array APD bias, overheat the local area, and cause an FPA failure. The short-related failure may occur in the array fabrication, FPA assembly, and later usage. In order to improve the FPA yield, uniformity, and reliability, the pixel input circuit was modified and effectively removed high dark current (shorted) APD pixels from the array to reduce the overall current draw of an APD array. Spectrolab has also introduced a feature to the APD detector array that reduces the optical crosstalk between pixels to <100ppm and reduces Dark Count Rate (DCR) at operating temperature to <10 kHz. External PDE on detectors alone, without a microlens, in excess of 40% has been demonstrated for reverse-illuminated arrays.

Parameter	Spec	Test conditions
Wavelength	1.06 µm	
Format	32x32	
Pixel Pitch	100 µm	
PDE	40%	240K, 4V overbias
Fill Factor	70%	
DCR	20 kHz	240K, 4V overbias
Cross talk	<1%	Optimized overbias for SCA
Pixel operability	99%	
RMS timing jitter	0.5 ns	

Table 1. Typical single photon counting focal plane array figures of merits for a 32x32 GM-APD array presently fabricated at Spectrolab.

All these parameters mentioned above must be optimized for the optimal sensor performance. In practice these parameters are not decoupled and often trade off against one another. Table 1 lists the most important FPA figures of merit that impact single photon counting sensor systems.

2. GM-APD DEVICE

In Geiger-mode operation the avalanche photodiodes are biased above the breakdown voltage, $\sim 2\text{-}4~\text{V}$, for a short period of time during which the sensor is active. Due to various carrier generation mechanisms in GM-APDs, there is some probability for the pixels to generate some output pulses even without any incident photons. This probability is characterized by dark count rate (DCR) and it represents the noise of GM-APD arrays. The maximum length of this active period is also determined by DCR. If a photon in the sensitive spectrum falls on the detector pixel in this period, it has a probability to be absorbed and multiplied in the GM-APD such that the amplitude of the result pulse is large enough to be converted into a digital pulse that can be registered by the following readout circuit. This probability is a figure of merit called photon detection efficiency (PDE).

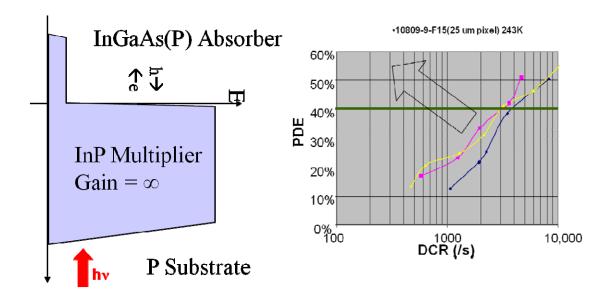


Figure 1 The electric field profile during operation Geiger-mode operation of an APD. The right panel shows PDE as function of DCR of three typical Spectrolab 25- μ m diameter Geiger-mode APD pixels of a 32x128 array. The data are measured at various overbias points and at 243 K.

In order to suppress the dark current generation in the narrow band absorber and achieve high gain in the multiplier, the InGaAsP/InP GM-APD separate absorption and gain and multiplication (SAGM) device structure is tuned for 1.06 µm operation and the electric field profile is shown in Figure 1. An InGaAsP alloy is used in the absorption layer for 1.06-µm photons. DCR increases with overbias and operating temperature. Because a higher overbias also yields a better PDE, for a better signal-to-noise ratio, the trade-off between PDE and DCR has to be carefully considered in the device operation point selection. The right panel plots the PDE against DCR of three 25-µm array pixels on a 32x128 GM APD array under different overbiases and at 243K. Grown on p-type InP substrates, the devices can achieve 40% PDE with 3~4 kHz DCR at the operation temperature. The device growth, process and characterization details have been reported previously^{2,3}.

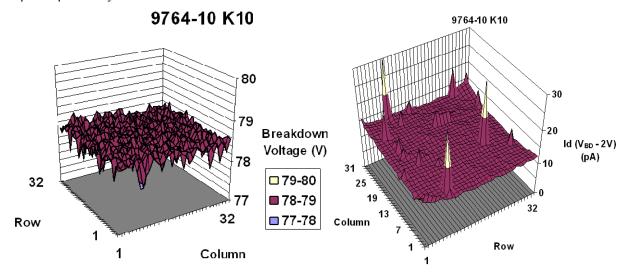


Figure 2 The breakdown voltage (Vbr) and dark current (Id) map of a typical 32x32 GmAPD array. The median Vbr is 78.5V and the standard deviation is 0.16V in the left panel. The

Due to the epitaxial non-uniformity, normally there is a breakdown voltage (Vbr) variation across the array. For most FPA applications, the common anode is biased with a single negative voltage, while the pixels are applied a positive pulse uniformly in the detection gate. Under this bias scheme, the overbias to each pixel will vary with the breakdown voltage and result in non-uniformity in DCR and PDE. For uniform performance, it is critical to minimize the Vbr variation across the array. Figure 2 shows the Vbr and Id maps of a typical 32x32 GmAPD array The standard deviation of Vbr is about 0.16 V, and that of Id is 3.7 pA.

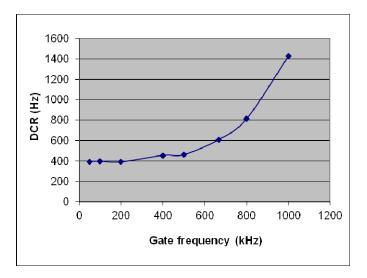


Figure 3 The afterpulsing induced DCR rise. It is measured at 243 K and 4 V overbias with a 30 μ m pixel and 20 ns gate. At the low gate frequencies, the DCR is about 400 Hz. The afterpulsing effect has little impact until about 500 kHz gate frequency, which is far beyond the normal camera frame rate.

In each avalanche event, a large amount of carriers are generated and transported through the APD junction. Some carriers will be trapped in the hetero-interfaces and trapping defects. As released later, they may trigger a false alarm and result in a higher dark count rate if the operation cycle is shorter than or comparable to the trap lifetime. One way to characterize this afterpulsing effect is to measure the DCR at different gate frequencies. Comparing to the planar APD structures, the mesa pixels in our array fully deplete the absorber and multiplier in the whole operation cycle and avoid the long trapping time of the photo- or avalanche-generated carriers in weak electric field. As shown in Figure 3, a typical 30 μ m pixel was measured at 243 K, 4 V overbias, and 20 ns gate. At the low gate frequencies, the DCR is about 400 Hz. The afterpulsing effect has little impact until the gate frequency is about 500 kHz, which is far beyond the normal camera frame rate.

With extensive capacities in MOVPE epitaxy, wafer process, and module packaging, Spectrolab vertically integrated the LADAR camera manufacturing in house. After the detector array fabrication, Indium bumps were evaporated on to the array pixels. The detector array was integrated with a ROIC of the same format by flip-chip bonding. A GaP microlens array was attached to the focal plane array for a better fill factor afterward.

Figure 4 shows the cross section of a 32x128 SCA. The FPA and the interposer were mounted on to a TE cooler and the assembly was soldered in a 160-pin grid array package. Later, the package was hermetically sealed with an AR-coated sapphire window.

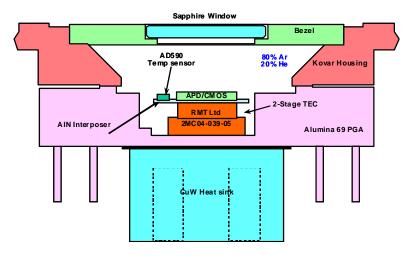


Figure 4. The cross section of a Spectrolab 32x128 SCA.

3. SPECTOLAB GEN-II 32X32 LADAR CAMERA

With the support from DARPA, a 32x32 readout-integrated circuit (ROIC), BFE316, was designed in 180-nm CMOS technology by Black Forest Engineering (BFE). Many improvements were incorporated in this design based on our experience with the previous versions. They include: short pixel protection, 4-bit APD array NUB correction, 7 μ s gate width with 0.5 ns resolution (or 1000 m range with 7.5 cm resolution), 435-ps whole system jitter, negligible chirp, 14V max overbias, and 300 mW power consumption⁴.

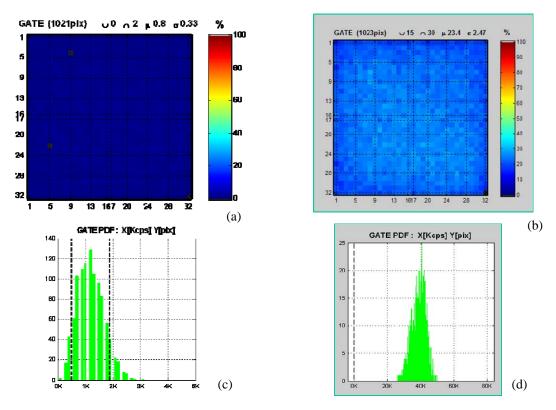


Figure 5 The DCR and PDE uniformity of an SCA without NUB correction. In the operation condition, (a) shows the DCR map with a mean of 0.8 kHz and 0.33 kHz in standard

deviation. (c) shows the corresponding DCR histogram. The PDE map shown in (b) is measured under the same operation condition with flood illumination. The mean count is 23.4 kHz with a standard deviation of 2.47 kHz.

Uniformity is a critical array performance. As shown in Figure 2, APD arrays have natural variations in breakdown voltage, which results in a performance variation across the array if it is biased under a uniform bias. A lot of effort has been invested in its improvement in semiconductor growth, device process, and SCA assembly. Figure 5 shows the DCR and PDE maps and histogram of an SCA at the operation condition without NUB correction. In the operation condition, (a) shows the DCR map with a mean of 0.8 kHz and 0.33 kHz in standard deviation. (c) shows the corresponding DCR histogram. The PDE map shown in (b) is measured under the same operation condition with flood illumination. The mean count is 23.4 kHz with a standard deviation of 2.47 kHz. Because of the difficulty of the illumination calibration, the absolute PDE value of each pixel is not available from this SCA measurement. In applications its value is estimated with the device performance as shown in Figure 1.

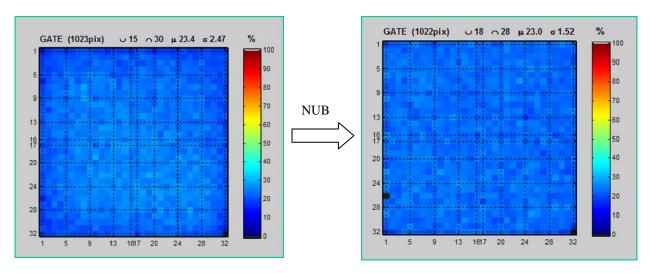


Figure 6. The PDE map of a 32x32 SCA operated at 242K and a frame rate of 12.5 kHz.

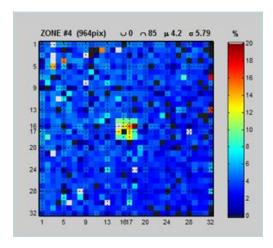


Figure 7 Enhanced sensitivity in a 4x4 region near the array center realized with the NUB function.

By introducing a 4-bit 2.5V in-pixel programmable NUB adjustment, even better performance uniformity can be achieved. Figure 6 demonstrated the NUB improvement in PDE of the same SCA shown in Figure

5. With a tuning range of $\pm 0.75 V$ in the individual pixel bias and the PDE map, the NUB function reduced the photo count standard deviation from 2.47 kHz to 1.52 kHz, while keeping the photo count mean nearly unchanged. This function also allows array windowing: increased or decreased sensitivity within a specified region. Figure 7 demonstrated a 4x4 region with enhanced sensitivity near the array center. Working like a digital optical aperture, this windowing function can avoid the interference and cross talk from other pixels out of the target area.

Short pixel protection is another important feature implemented in this ROIC. With the immunity to sparse shorted detectors normally in APD arrays, this protection in the front end circuit greatly improves manufacturability and stability of FPAs.

The ROIC power consumption is reduced to 300 mW, about 50% of the previous version. Because the FPA is at the top of several stages of TECs, which consume most of the power in a LADAR camera, the nearly halved ROIC power consumption dramatically reduced the camera power load and heatsink weight.

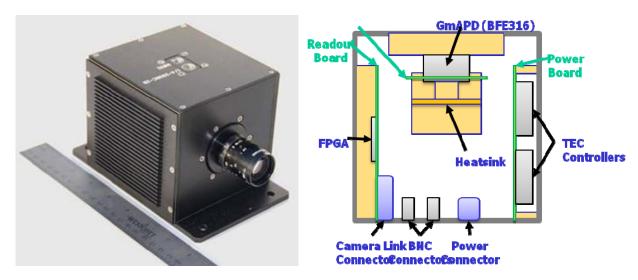


Figure 8 Spectrolab Gen-II 32x32 GmAPD LADAR camera.

With the readout boards developed at Boeing DES, the Gen-II 32x32 LADAR camera was integrated with the SCA based on the BFE 316 ROIC. Figure 8 illustrates this camera and its internal structure. With GM APDs, the single photon level sensitivity greatly extends the LADAR system range, reduces the laser power requirements and the lens size, and minimizes the SWAP of the overall system. In Spectrolab Gen-II 32x32 LADAR camera, the detection gate can be kept on for about $7 \mu s$, which provides a range window of about 1 km long. Additional range coverage can be realized by inserting a delay between the laser pulse and gate turn-on and shifting the range window until reaching the maximum range allowed by the laser power and the array sensitivity. With a proper delay, the laser signal will fall in the detection gate and fire pixels in the array. The following circuits of the fired pixels will latch the timing of the detection. At the end of the gate, the whole array will be scanned and the timing information will be read out.

The ranging resolution of a LADAR camera is determined by its overall jitter and chirp performance. The temporal histogram of our Gen-II 32x32 camera to a synchronized 90 ps laser pulse is shown in the top panel of Figure 9. The bin size is 500 ps, and the laser pulse width is included in this measurement. By calculating the standard deviation of this temporal distribution, we can get the overall system RMS jitter of 418 ps, which includes the laser pulse width and jitters in the trigger loop, APDs, ROIC, and readout boards and is shorter than the 500ps bin size. With a new clock design, the chirp in this camera is negligible. As shown in the bottom panel of Figure 9, by increasing the delay of a laser pulse 500 ns each time, we have got a series of evenly spaced signals in the detection gate.

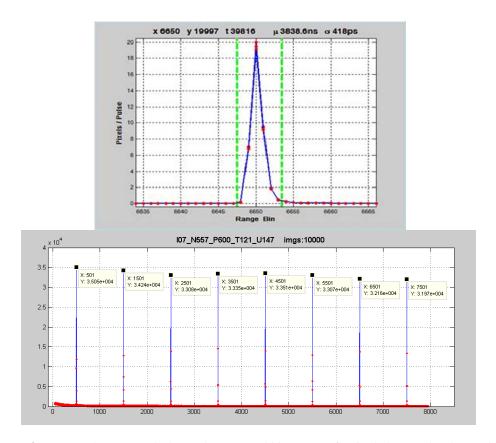


Figure 9 The top graph shows the temporal histogram of a single laser pulse detected by a Spectrolab Gen-II 32x32 LADAR camera. The overall system RMS jitter is about 418 ps. The bottom panel shows a series of signals detected from laser pulses delayed by 500 ns from each other.

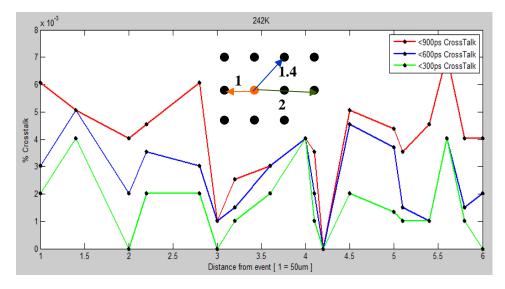


Figure 10 The crosstalk probability as a function of the distance to the initiating pixel. The pixel distance used in the x-axis is illustrated in the inserted diagram.

Because the GM APDs are biased beyond the breakdown voltage and they are extremely sensitive to single photon level in the detection gate, any secondary emission from an avalanche event in an array can propagate to the neighbor pixels and trigger false detections. If this crosstalk probability is not well controlled, the chain reaction can turn on the whole array with a single input photon and severely degrade the LADAR image quality. As the pixel pitch reduced from $100~\mu m$ of the previously reported 32x32~GM APD arrays to $50~\mu m$, the crosstalk is an even greater concern. Several layers of proprietary isolation schemes were employed in the array design and process.

Figure 10 shows the 50 μ m pitch arrays demonstrating a crosstalk level about 60 ppm under the operation condition, which is about one order of magnitude less than that of the 100 μ m pitch 32x32 arrays we reported before⁴.

4. LADAR SYSTEM FIELD TEST

Boeing DES developed a compact LADAR system based on the Spectrolab Gen-II 32x32 LADAR camera demonstrated above. The total weight is less than 20 lbs with the lens and scanning mechanics. The size is 990 cu. in., and the power consumption is less than 325W. With data reduction from hundreds of MB/s raw data to about 20 MB/s, this compact LADAR system gives the user rapid 3D imagery to access the mapped area quickly and accurately. At 1000 ft AGL, this system can cover 0.5 sq. mi/hour with a range resolution of 6 inches and a GSD of 5 inches. This system has been demonstrated on various ground and aerial platforms.



Figure 11 The compact LADAR system integrated with the Spectrolab Gen-II LADAR camera by Boeing DES.

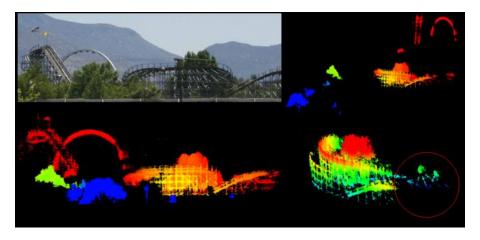


Figure 12 Final static test of amusement park about 1000 ft away from Boeing-SVS office. The rotated image shows small Ferriswheel hidden in the view from the LADAR location.

As shown in Figure 12, the final ground static test demonstrated preparedness and highlights benefit of photon counting LADAR. The roller coaster clearly shows the depth even though the image was taken from one position. The rotated image shows small Ferris wheel hidden in the view from the LADAR location.

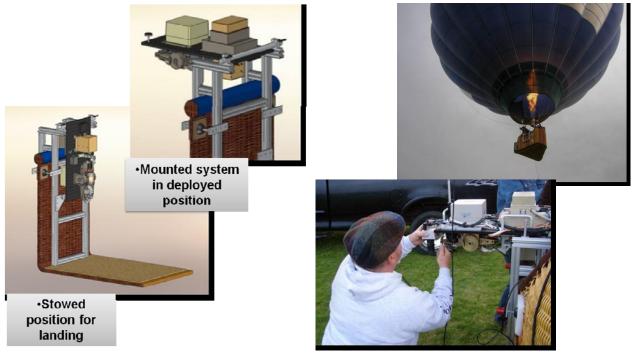


Figure 13 Final LADAR test system on a hot air balloon integrated with GPS/INU and battery power.

Figure 14 Hot air balloon demo on May 18, 2010.

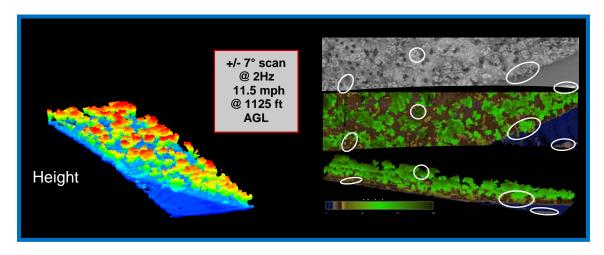


Figure 15 3D images of forested areas and the Rio Grande near Albuquerque, NM generated by a compact Boeing DES LADAR system mounted on a hot air balloon.

Figure 13 shows the LADAR test system on a hot air balloon integrated with GPS/INU and battery power. Two different mounting positions are prepared for uncontrolled balloon landing and mounting to the wicker basket. The integration went smoothly with only minor issues with the GPS ground station at start-up. There are no hardware issues in the flight. The LADAR system was demonstrated on a hot air balloon platform. The flight took a little more than one hour, and the system collected over 30 minutes of data.

Figure 15 shows the 3D images of forested areas and the Rio Grande near Albuquerque, NM passed by in this flight. The images were taken at 1125 ft AGL and 11.5 mph, and the LADAR system scanned in a range of $\pm 7^{\circ}$ at 2 Hz.

5. 32X128 GMAPD SCA

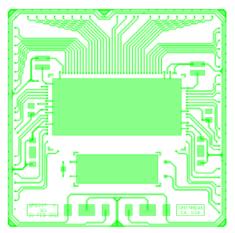


Figure 16 The interposer for 32x128 GmAPD SCA.

To answer the challenges of SWIR LADAR applications, Spectrolab, teamed with Boeing DES and Black Forest Engineering, developed a 32x128 GM-APD SCA as reported previosly³. Based on this effort and the experience in our 32x32 SCAs, the commercial 32x128 SCA has incorporated many new features, including full Cameralink control and output, 50 kHz frame rate at all gate width, 1000 m max range with programmable $0.3 \sim 1.0$ ns bins, 14-bit timing resolution, short pixel protection, and 4-bit NUB correction. Although the size is four times of that of the 32x32 SCAs, the power consumption is kept at 300 mW, the same as the latest Gen-II 32x32 SCA. As shown in the interposer drawing in Figure 16, the FPA is located in the center of the SCA. Similar to the 32x32 SCAs, the final assembly will be hermetically sealed in a 160-pin PGA package with the same size and footprint. The cross section of the final SCA is shown in Figure 4. Based on this SCA development, the next generation GmAPD 32x128 LADAR camera will be demonstrated soon.

6. SUMMARY

For the SWIR LADAR applications, Spectrolab developed high performance 32x32 and 32x128 GmAPD image arrays and SCAs with 3-4 khz DCR at 40% PDE, 60 ppm crosstalk at 50 μ m pitch, 0.16V Vbr standard deviation, and no afterpulsing until 500 kHz frame rate. With many new features for manufacturability, operation stability and reliability, and application convenience, the Gen-II 32x32 GmAPD LADAR system has been demonstrated low SWAP footprint on various platforms with excellent resolution. For aerial mapping and reconnaissance application, a new 32x128 SCA and LADAR camera are under development.

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